

MAGNETIC CIRCUIT DESIGN

1. MAGNETIC CIRCUIT ANALYSIS

1-1. Basic calculation method

The basic calculation method of a magnetic circuit is the same as is used in a basic electrical analysis using Ohm's Law.

The total magnetic flux ϕ (analogous to electric current), magnetomotive force F (analogous to voltage), and magnetic reluctance R (analogous to electrical resistance) are related as shown in below.

$$\text{Total magnetic flux } \phi = \frac{\text{Magnetomotive force } F}{\text{Magnetic reluctance } R} \dots\dots\dots(1)$$

In magnetic circuit calculations, it is more common to use the magnetic permeance P , which is the reciprocal to reluctance R . Using permeance instead of reluctance, the total flux equation is changed as shown in below.

$$\text{Total magnetic flux } \phi = \text{Magnetomotive force } F \cdot \text{Permeance } P \dots\dots\dots(2)$$

The permeance P is a function of the magnetic circuit length L , magnetic circuit cross sectional area A , and magnetic permeability μ .

$$\text{Permeance } P = \frac{\text{Permeability } \mu \cdot \text{Cross sectional area } A}{\text{Magnetic circuit length } L} \dots\dots\dots(3)$$

This means that when the length is shorter and the magnet area and permeability are greater, the permeance is greater. Conversely the reluctance is reduced.

The total magnetic circuit permeance P_t is considered as a total of any air gap permeance P_g defined as a reciprocal of magnetic reluctance in the air gap, and the sum of leakage permeance P_f , defined as the reciprocal of the magnetic reluctance of the leakage flux paths at the poles.

$$P_t = P_g + P_f = P_g + P_{f1} + P_{f2} + \dots\dots\dots + P_{fn} \dots\dots\dots(4)$$

1-2. Magnetomotive force loss coefficient f

The magnetomotive force loss coefficient f is the ratio of the total magnetomotive force F_t and the magnetomotive force in the air gap F_g for a given magnetic circuit.

$$\text{Magnetomotive force loss coefficient } f = \frac{F_t}{F_g} \dots\dots\dots(5)$$

The total magnetomotive force F_t in the magnetic circuit is determined as the product of the magnetic field strength H_d at the operating point, and the length of the magnet L_m .

The magnetomotive force in the air gap F_g is given as the product of the magnetic field strength of the air gap $H_g (= B_g)$, and the length of the air gap L_g .

Thus equation (5) becomes as follows.

$$\text{Magnetomotive force loss coefficient } f = \frac{H_d \cdot L_m}{H_g \cdot L_g} = \frac{H_d \cdot L_m}{B_g \cdot L_g} \dots\dots\dots(6)$$

1-3. Leakage coefficient σ

The leakage coefficient is the ratio of the total magnetic flux ϕ_t generated from the magnet in a given circuit and the flux found in the air gap ϕ_g .

$$\text{Leakage coefficient } \sigma = \frac{\phi_t}{\phi_g} \dots\dots\dots(7)$$

The total magnetic flux generated in a magnet ϕ_t is given as the accumulation of flux density at the operating point B_d over the cross sectional area of the magnet A_m .

And the magnetic flux in the air gap ϕ_g is given as the accumulation of flux density B_g over the area of the air gap A_g .

Equation (7) now becomes:

$$\text{Leakage coefficient } \sigma = \frac{B_d \cdot A_m}{B_g \cdot A_g} \dots\dots\dots(8)$$

From equation (2), we can rewrite equation (7) as follows:

$$\text{Leakage coefficient } \sigma = \frac{F_t \cdot P_t}{F_g \cdot P_g} \dots\dots\dots(9)$$

Generally the magnetomotive force loss coefficient f is approximately 1, so equation (9) becomes:

$$\text{Leakage coefficient } \sigma \approx \frac{P_g + P_f}{P_g} = 1 + \frac{P_f}{P_g}$$

Splitting P_f in each part of the magnetic leakage path, the leakage coefficient σ can be written as equation (10).

$$\text{Leakage coefficient } \sigma \approx 1 + \frac{P_{f1} + P_{f2} + \dots\dots\dots + P_{fn}}{P_g} \dots\dots\dots(10)$$

1-4. Permeance coefficient P_c

The permeance coefficient is used to design a permanent magnet application with a B-H curve. This is defined as the ratio of flux density B_d and magnetic field strength H_d of the operating point, and equation (11) becomes:

$$\text{Permeance coefficient } P_c = \frac{B_d}{H_d} \dots\dots\dots(11)$$

The relationship is shown in figure 1 below:

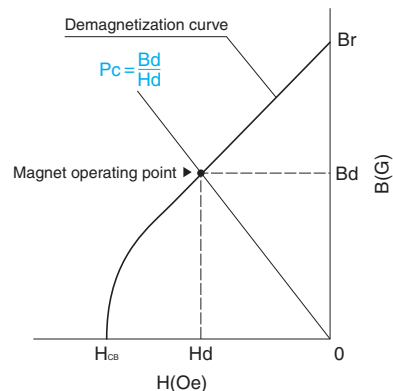


Fig. 1

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1-4-a. Permeance coefficient in a magnetic circuit

The permeance coefficient in a magnetic circuit can be rewritten from equations (6) and (8), and then they are placed back into equation (11).

Equation (6) therefore,

$$Hd = \frac{Bg \cdot Lg}{Lm} f \quad \text{.....(12)}$$

Equation (8) therefore,

$$Bd = \frac{Bg \cdot Ag}{Am} \sigma \quad \text{.....(13)}$$

Then,

$$Pc = \frac{Lm \cdot Ag}{Am \cdot Lg} \cdot \frac{\sigma}{f} \quad \text{.....(14)}$$

1-4-b. Permeance coefficient of a magnet in an open circuit

When a magnet is used in an open circuit, the permeance coefficient is greatly affected by the magnet shape. Therefore it is extremely difficult to calculate the exact value.

The following figures of the cylindrical type magnet (the relationship graph of the dimension ratio and permeance coefficient) can be used to approximate the permeance.

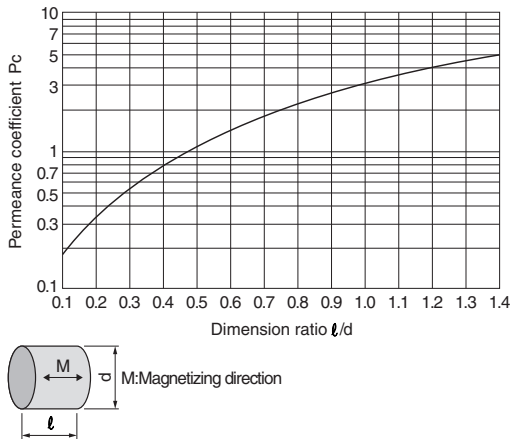


Fig. 2

The approximate equation is as follows:

$$Pc = 1.35 \left(\frac{l}{d} \right) \left[\sqrt{1 + \left(\frac{l}{d} \right)^2} + \left(\frac{l}{d} \right) \right] \quad \text{.....(15)}$$

1-5. Calculation of necessary magnet length L_m , cross sectional area A_m and volume V_m

Equation (6) therefore,

$$Lm = \frac{Bg \cdot Lg}{Hd} f \quad \text{.....(16)}$$

Equation (8) therefore,

$$Am = \frac{Bg \cdot Ag}{Bd} \sigma \quad \text{.....(17)}$$

If Hd , Bd , Ag , Lg , f and σ are provided, Lm and Am can be found using equations (16) and (17). The required magnet volume V_m is calculated as follows:

$$Vm = Am \cdot Lm = \frac{Bg^2 \cdot Ag \cdot Lg \cdot \sigma \cdot f}{Bd \cdot Hd} \quad \text{.....(18)}$$

The necessary magnet volume is inversely proportional to the energy product at a particular operating point.

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1-6. Method to determine the leakage coefficient σ , and magnetomotive force loss coefficient f

1-6-a. Experimental method

With a search coil

1. A search coil was winding around the magnet to measure the gap flux.
2. Bd is calculated from total flux ϕ_g /magnet sectional area A_m .
3. Determine Hd using the B-H demagnetization curve.
4. Measure Bg , Ag , Lg , A_m and L_m .
5. Calculate f and σ using equations (6) and (8).

Without a search coil

1. Hypothesize f (generally 1.0 to 1.2).
2. Measure L_m , Bg and Lg .
3. Calculate Hd is using equation (12).
4. Determine Bd from the B-H demagnetization curve.
5. Measure A_m and Ag , and calculate σ using equation(8).

1-6-b. Calculating leakage coefficient σ

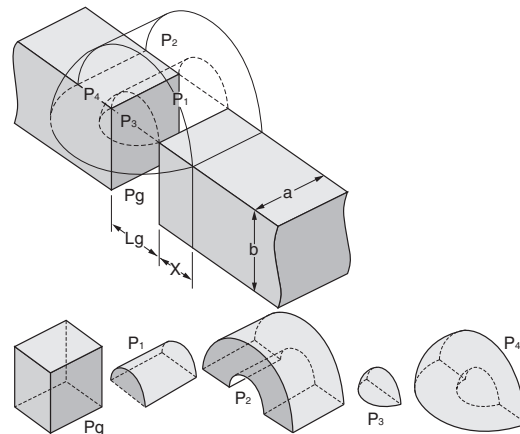
The leakage coefficient is calculated as follows:

Determine P_g and P_{f1} to P_{fn} by calculation (refer to the example shown in below) and calculate σ using equation (10).

$$\text{Equation (10): } \sigma \doteq 1 + \frac{P_{f1} + P_{f2} + \dots + P_{fn}}{P_g}$$

(Example)

An example of the calculated permeance for a magnetic circuit is shown in below.



1. Permeance of the space gap

$$P_g = \mu_0 \frac{a \cdot b}{L_g}$$

2. Permeance of half cylinder portion

$$P_1 = 0.264\mu_0 \cdot a$$

3. Permeance of the hollow half cylinder portion

$$P_2 \approx \frac{0.64\mu_0 \cdot a}{(L_g/x+1)} \text{ or } P_2 = \frac{\mu_0 \cdot a}{\pi} \ln\left(1 + \frac{2X}{L_g}\right) \quad [L_g < 3x]$$

4. Permeance of the quadsphere portion

$$P_3 = 0.077\mu_0 \cdot L_g$$

5. Permeance of the hollow quadsphere portion

$$P_4 = \frac{\mu_0 \cdot x}{4}$$

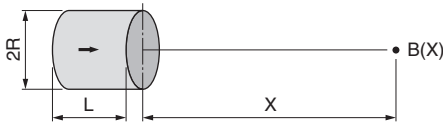
By combining the permeance of these respective portions, the approximate leakage coefficient in the circuit can be found with using equation (10).

2. EQUATION BY DETERMINING THE MAGNETIC FLUX DENSITY AT A DISTANCE X FROM THE MAGNET AT THE CENTER LINE, B(X)

When the operating point Bd is above the knee point of the B-H curve, the magnetic field distribution on the outside of the magnet, assuming the space of length X and the same cross section shape with permeability equal to the magnet, is similar to the magnetic field generated by the closed circuit current on the outer surface.

The equations for three typical magnet shapes and some applications are shown in below. (These formulas can be used for ferrite or rare earth magnets.)

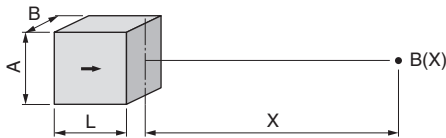
2-1. Cylindrical shaped



$$B(X) = \frac{Br}{2} \left[\frac{L+X}{\sqrt{R^2+(L+X)^2}} - \frac{X}{\sqrt{R^2+X^2}} \right]$$

Br: Residual flux density of the magnet
X: Distance from the surface of the magnet

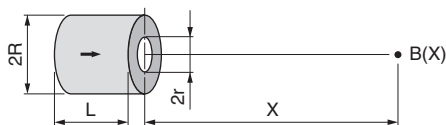
2-2. Block shaped



$$B(X) = \frac{Br}{\pi} \left[\tan^{-1} \frac{AB}{2X\sqrt{4X^2+A^2+B^2}} - \tan^{-1} \frac{AB}{2(L+X)\sqrt{4(L+X)^2+A^2+B^2}} \right]$$

Br: Residual flux density of the magnet
X: Distance from the surface of the magnet
The angle is radian.

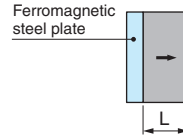
2-3. Ring shaped



$$B(X) = \frac{Br}{2} \left[\left(\frac{L+X}{\sqrt{R^2+(L+X)^2}} - \frac{L+X}{\sqrt{r^2+(L+X)^2}} \right) - \left(\frac{X}{\sqrt{R^2+X^2}} - \frac{X}{\sqrt{r^2+X^2}} \right) \right]$$

Br: Residual flux density of the magnet
X: Distance from the surface of the magnet

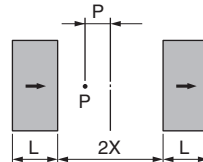
2-4. Ferromagnetic steel plate on the back of the magnet



The steel plate must have sufficient thickness that will be not to become saturated.

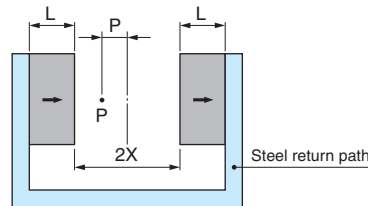
Substitute $2L$ for L in equations of 2-1., 2-2. and 2-3. above.

2-5. Calculating B(X) of same shaped two magnets when spaced at a distance of 2X



If two magnets are used as shown in above, $B(X)$ at the center of the air gap is doubled based on equations 2-1., 2-2. and 2-3.. The $B(X)$ at point P in the air gap is found using the same equations as above, except the sum of $B(X-P)$ and $B(X+P)$, where $X+P$ and $X-P$ are substituted into the equations for X .

2-6. Same shaped two magnets with a steel return path



Substitute $2L$ for L in the equations derived in the above 2-5..

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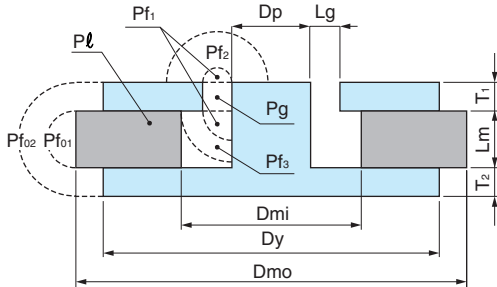
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3. CALCULATING THE AIR GAP FLUX DENSITY FOR A LOUD-SPEAKER MAGNETIC CIRCUIT

3-1. Calculation method for outer magnet type

Assume the permeance can be found based on the leakage flux path shown in the figure below.

The gap permeance P_g and the leakage permeance P_f in each part can be found as follows.



D_p : Center pole diameter
 T_1 : Top plate thickness
 D_{mo} : Outside radius of magnet
 L_m : Magnet thickness
 A_g : Gap cross sectional area
 D_y : Plate diameter
 T_2 : Bottom plate thickness
 D_{mi} : Inside radius of magnet
 L_g : Gap length

$$P_g = \mu_0 A_g / L_g = \pi \mu_0 (D_p + L_g) T_1 / L_g$$

$$P_{f1} = 0.264 \pi \mu_0 (D_p + L_g)$$

$$P_{f2} = \mu_0 (D_p + L_g) \ell n(1 + D_p / L_g)$$

$$P_{f3} = 2 \mu_0 (D_p + L_g) \ell n\{1 + (D_{mi} - D_p - 2L_g) / L_g\}$$

$$P_{f01} \doteq \mu_0 \cdot 0.264 \pi \cdot D_{mo} + 0.308 L_m$$

$$P_{f02} \doteq \mu_0 D_{mo} \cdot \ell n\{1 + (T_1 + T_2) / L_m\} + (T_1 + T_2)$$

$$P_L = \pi \mu_0 \sqrt{(D_{mo} + D_{mi}) L_m} / 2$$

We can calculate the total permeance P_t of the whole circuit from the equations shown in above.

$$P_t = P_g + 3P_{f1} + P_{f2} + P_{f3} + P_{f01} + P_{f02} + P_L$$

From the previous equation, we can derive an equation for the leakage coefficient.

$$\sigma = P_t / P_g$$

Then,

$$P_c = \frac{L_m}{A_m} \cdot P_g \cdot \frac{\sigma}{f} = \frac{L_m}{A_m} \cdot \frac{P_t}{f}$$

A_m : Cross sectional area of magnet

f : Magnetomotive loss coefficient $\doteq 1.1$

The flux density at the magnet operating point, B_d :

$$B_d = \frac{B_r}{1 + \mu_{rec} / P_c}$$

μ_{rec} : Recoil permeability $\doteq 1.7$

The air gap magnetic flux density, B_g :

$$B_g = \frac{A_m \cdot B_d}{A_g} \cdot \frac{1}{\sigma}$$

The bottom plate must have sufficient thickness T_2 not to become saturated. This is determined as follows:

The leakage coefficient as viewed from the bottom plate is:

$$\sigma_p = (P_g + 3P_{f1} + P_{f2} + P_{f3}) / P_g$$

The total magnetic flux passing through the bottom plate is:

$$\phi_p = B_g \cdot A_g \cdot \sigma_p$$

Assuming the saturated flux density of the bottom plate is 15kG, the thickness of the bottom plate T_2 should be:

$$15000 \pi \cdot D_p \cdot T_2 \geq \phi_p$$

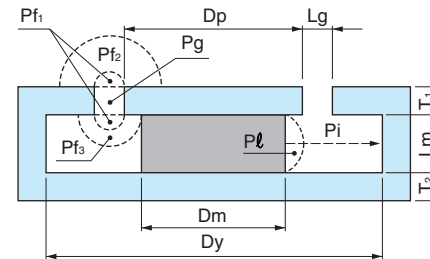
Then,

$$T_2 \geq \phi_p / 15000 \pi \cdot D_p$$

3-2. Calculation method for inner magnet type

Assume the permeance can be found based on the flux leakage path shown in figure below in the same way as for the outer magnet type.

The gap permeance P_g and the leakage permeance P_f in each part can be found as follows:



D_p : Center pole diameter
 T_1 : Top plate thickness
 D_m : Diameter of magnet
 A_g : Gap cross sectional area
 D_y : Inside plate diameter
 T_2 : Bottom plate thickness
 L_m : Magnet thickness
 L_g : Gap length

$$P_g = \mu_0 A_g / L_g = \pi \mu_0 (D_p + L_g) T_1 / L_g$$

$$P_{f1} = 0.264 \pi \mu_0 (D_p + L_g)$$

$$P_{f2} = \mu_0 (D_p + L_g) \ell n(1 + D_p / L_g)$$

$$P_{f3} = \mu_0 (D_p + L_g) \ell n\{1 + (D_p - D_m) / L_g\}$$

$$P_L = \pi \mu_0 \sqrt{D_m \cdot L_m} / 2 \quad [P_L > P_i]$$

or

$$P_i = \pi \mu_0 L_m (D_m + D_y) / 4 (D_y - D_m) \quad [P_i > P_L]$$

Then the total permeance P_t is:

$$P_t = P_g + 2P_{f1} + P_{f2} + P_{f3} + P'$$

P' is P_L or P_i , whichever is larger.

The calculation process of the leakage coefficient σ , permeance coefficient P_c , flux density at the operating point B_d , and air gap flux density B_g is the same as was shown for the outer magnet type.

The plate thickness T_1 and T_2 required for the plate not to become saturated is shown in below, and the calculation process is the same as for the outer magnet type.

The leakage coefficient is:

$$\sigma_p = (P_g + 2P_{f1} + P_{f2} + P_{f3} + P_L \text{ or } P_i) / P_g$$

The total magnetic flux ϕ_p through the plate is:

$$\phi_p = B_g \cdot A_g \cdot \sigma_p$$

Assuming the saturated flux density of the plate is 15kG, T_1 and T_2 should be:

$$T_1, T_2 \geq \phi_p / 15000 \pi \cdot D_m$$

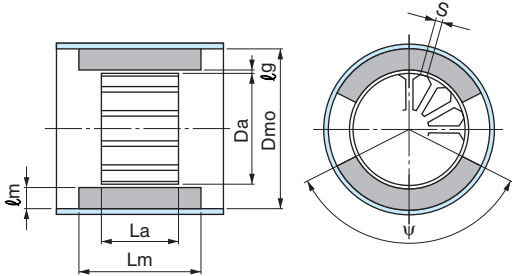
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4. CALCULATION OF THE EFFECTIVE MAGNETIC FLUX IN A MOTOR

There are many variables that change the magnetic circuit of a motor greatly, such as the number of rotor slots, slot geometry, motor housing thickness, etc.

Therefore the following calculation shows only the basic concepts



Dmo: Outside radius of magnet
Lm: Magnet axial length
 ψ : Magnet angle(radians)
 l_m : Magnet pole thickness

Da: Armature diameter
La: Armature length
S: Slot width
 l_g : Air gap length

1) Cross sectional area A_m of the magnet is as follows:

$$A_m = \frac{1}{2} \psi L_m (D_{mo} - l_m) = \frac{1}{2P} \alpha_i \pi L_m (D_{mo} - l_m)$$

$\alpha_i = p \psi / \pi$: An arc rate of a pole, P: The binary number of poles

2) Magnet equivalent value shape ratio l/d

$$\frac{l}{d} = \frac{l_m}{\sqrt{4A_m/\pi}}$$

3) Leakage permeance coefficient P_i and P_L

• Permeance coefficient of the magnet alone P_i

$$P_i = \zeta \left(\frac{l}{d} \right) \left[\sqrt{1 + \left(\frac{l}{d} \right)^2} + \frac{l}{d} \right]; \zeta = \frac{1}{5} (12 - p)$$

• Permeance coefficient of the magnet alone (including the yoke) P_L

$$P_L = \xi \left(\frac{2l}{d} \right) \left[\sqrt{1 + \left(\frac{2l}{d} \right)^2} + \frac{2l}{d} \right]; \xi = 1.3 \sim 1.4$$

4) Carter coefficient k_c

$$k_c = \frac{t_s}{t_s - \gamma_c l_g}; \gamma_c = \frac{(S/l_g)^2}{5 + (S/l_g)}; t_s = \frac{\pi D_a}{S n}$$

Where,

t_s : tooth pitch

$S n$: Number of slots

5) Effective permeance coefficient P_u

$$P_u = \frac{l_m}{A_m} \cdot \frac{A_g}{k_c l_g}; A_g = \frac{1}{2} \psi k_i L_a (D_a + l_g); k_i \approx 0.97$$

Where,

A_g : Air gap cross sectional area

k_i : Lamination factor in the armature

6) Leakage coefficient σ

$$\sigma = 1 + P_i/P_u, \quad \sigma P_u = P_u + P_i$$

7) Magnetic flux at the pole (Effective magnetic flux) ϕ_g

$$\phi_g = 0.95 B_r \frac{A_m \cdot P_u}{\mu_{rec} + \sigma P_u}$$

Where,

μ_{rec} : Recoil permeability

The above calculation method gives no special consideration to saturation of the magnetic circuit. In reality there can be varying degrees of saturation in the motor housing and rotor.

When comparing ϕ_g calculated above with the actually measured value, if the actually measured value is smaller than the calculated value, the motor circuit can be considered to have some saturation in one or more of the components.

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CONVERSION TABLE OF SI AND CGS UNITS

View in conversion table:

If the conversion ratio at the left of ► sign is multiplied by the column of SI unit system, the value becomes in CGS system.

If the conversion ratio at the right of ◄ sign is multiplied by the column of CGS unit system, the value becomes in SI system.

Terms	Symbol	SI units		Conversion(Multiplication)		CGS units	
		Unit's name	Symbol	SI to CGS	CGS to SI	Unit's name	Symbol
Magnetic flux	ϕ	Weber	Wb	10^8 ►	◄ 10^{-8}	Maxwell	Mx
Flux density	B	Tesla	T	10^4 ►	◄ 10^{-4}	Gauss	G
Magnetic field strength	H	Ampere/meter	A/m	$4\pi \times 10^{-3}$ ►	◄ $10^3/4\pi$	Oersted	Oe
Intensity of magnetization	M	Ampere/meter	A/m	10^{-3} ►	◄ 10^3	Gauss	G
Magnetic polarization	J	Tesla	T	$10^4/4\pi$ ►	◄ $4\pi \times 10^{-4}$	Gauss	G
Magnetomotive force	Fm	Ampere	A	$4\pi \times 10^{-1}$ ►	◄ $10/4\pi$	Gilbert	Gi
Force	F	Newton	N	10^5 ►	◄ 10^{-5}	Dyne	dyn
Permeability	μ	Henry/meter	H/m	$10^7/4\pi$ ►	◄ $4\pi \times 10^{-7}$	—	
Permeability, vacuum	μ_0	$4\pi \times 10^{-7}$ Henry/meter	H/m			1	
Reluctance	Rm	1/Henry	H ⁻¹	$4\pi \times 10^{-9}$ ►	◄ $10^9/4\pi$	Gilbert/Maxwell	Gi/Mx
Permeance	P	Henry	H	$10^9/4\pi$ ►	◄ $4\pi \times 10^{-9}$	Maxwell/Gilbert	Mx/Gi
Energy product	BH	Joule/metric ³	J/m ³	$4\pi \times 10$ ►	◄ $10^{-1}/4\pi$	Gauss • Oersted	G • Oe
				10 ►	◄ 10^{-1}	erg/cm ³	erg/cm ³

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